The simulation of oscillating wave surge converters using a Boussinesq model; Wave disturbances around an array

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ABSTRACT

A new method for simulating a frequency independent absorption within DHI’s Mike 21 Boussinesq wave (BW) model is presented. This provides an increase in the accuracy of the simulation of wave processes around a hypothetical WEC array. Multiple monochromatic wave simulations are combined to represent a wave spectrum. Wave device characteristics are then simulated using porosity layers. A frequency dependent porosity for each device is then applied based on data taken from an experimental study. This method is tested for nearshore shallow water devices where the wave energy disturbance is quantified for flat and varying domain gradients.

KEY WORDS: Boussinesq wave model, Wave energy modelling, Frequency dependent absorption, Wave-device interactions, Coastal impacts

INTRODUCTION

Over recent years, interest in environmental impacts has driven a more sophisticated treatment of WECs within numerical wave models. Considering individual devices within arrays, rather than solid blocks, has led to a more accurate and detailed description of their effect on the wave climate (Venugopal, Smith 2007, Smith, Pearce & Millar 2012, Greenwood et al. 2013).

Popular simulation methods are reviewed in (Folley et al. 2012). Wave resource models conducted using spectral wave software, which provide good representation of wave propagation in coastal waters but lack accurate diffraction terms, should be supplemented by other simulation types to gather more information around wave-device disturbances. The use of Boussinesq/Mild slope models provides a much higher temporal and spatial detail without applying stringent domain size restraints as seen with CFD. Boussinesq models also include an explicit numerical representation of diffraction, allowing an improved simulation behind structures when compare to a spectral wave model.

Boussinesq models have been applied to the simulation of many device types in shallow water environments. Solid open walled structures have been used to simulate the presence of oscillating water columns within the BW model (Venugopal, Bryden & Wallace 2010). While allowing no transmission of energy, the method provides a reasonable approximation for nearshore solid structures. The use of a solid structure is less justifiable to simulate floating or energy permeable structures. Large floating Wave Dragon devices were simulated using a partial transmission of wave energy through a combination of porosity and sponge layers (Torch P. et al. 2010, Beels et al. 2010). The arms were considered to be permeable, while a sponge layer prevented the transfer of energy through the main body. The detailed analysis of the down-wave results agreed well with experimental data, but largely neglected up-wave effects. In this case, a frequency-independent treatment was justified due to the 0% energy transmission through the body of the device.

Non-terminal devices can be simulated using porosity structures within a BW model. The up and down wave disturbance from a floating hinge type absorber was considered by (Angelelli, Zanuttigh 2012, Angelelli, Zanuttigh & Kofoed 2012). Reasonable agreement with experimental datasets led to the conclusion that a properly calibrated porosity layer provides a good representation of reflected and transmitted waves around a device.

While previous methods have shown reasonable results in simulating specific WECs, device absorption has generally been assumed to be independent of frequency. This study presents a method where a device’s frequency response is included. This is conducted for a small array, and the spatial variation of wave disturbance is analysed in detail.

METHODOLGY

The simulation of the propagation of waves was conducted in DHI’s MIKE 21 BW (DHI 2012). This software simulates the 2-dimensional propagating waves in shallow water based on Boussinesq formulations (Madsen, Sorensen 1992). The BW model has primarily been used for the simulation of waves of around coastal structures such as harbours and breakwaters. A flap-like nearshore OWSC (Oscillating Wave Surge Converter) was chosen as a frequency-dependent case study. To quantify the change in wave energy around the devices each test case requires two complete models where the undisturbed and the wave-device interactions are simulated. The results of this model are then compared for each element, quantifying the relative change across the whole of the domain.
Frequency Dependent Absorption

Porous structures have been shown to offer a reasonable representation of an energy permeable device. This provides a broad bandwidth of absorption over all frequencies. To better represent an operational device within the BW model a basic frequency dependant absorption can be applied. This uses frequency dependent porosity structures for device simulations, where monochromatic boundary conditions are used. This is achieved by breaking down the incident wave spectrum into frequency bins. Each frequency bin is associated with a monochromatic wave with the same frequency, and amplitude chosen so that the wave’s energy is equal to the energy within the bin. Each monochromatic wave is propagated through the model separately, with the absorption altered to match the device’s frequency response. The transmitted spectrum is ultimately obtained by combining these monochromatic components. Note that the amplitudes of the individual waves depend on the frequency discretisation, which should be carefully considered to ensure that the individual waves are handled realistically by the BW model.

An experimental frequency based power capture, obtained by (Clabby et al. 2012) using a 1:20 scale Aquamarine Power Oyster device for 40 sea states, was adapted to provide a suitable representation of an OWSC. Fig. 1 shows the average frequency dependent power capture for the experimental tests. The power capture, originally normalised on the y-axis, has been de-normalised to adhere to a peak value of 0.2, deemed to be an appropriate peak absorption. A smoothed version of the power capture curve is plotted that reduces variance for higher frequency regions. The frequency dependant absorption was separated into 15 frequency intervals, ranging from 0.06Hz to 0.25Hz with a ∆f of 0.01333Hz. The location of each frequency bin is indicated in Fig. 1, where the $C_p$ (power capture) coefficient is indicated. The porosity value for each frequency interval was then given by $1-C_p$. This relationship was based on pre-experimental trials where the porosity structure dimensions and behaviour with respect to the incident wave conditions were tested.

Model Set-up

The two-dimensional (2D) wave model algorithm was applied to a hypothetical domain with a constant depth of 10m. A 2m regular grid resolution was combined with a time step of 0.15s to produce a Courant number of 0.743. The domain covers an area of 700 x 750 cells (1400m x 1500m). The simulations were run for a time period of 1800s. An internal wave generation line wave used as the driving boundary. The wave conditions were based on a JONSWAP spectrum with an $H_s = 2m$ and $T = 10s$ with the following parameters $\gamma=3.3$, $\sigma_a=0.07$ and $\sigma_b=0.09$. A sponge layer 50 cells thick was applied at either end of the domain to prevent internal wave reflections. Wave breaking and eddy viscosity were enabled where the where the type of roller celerity was set to type 1 and all other terms were kept at their default values.

A small array was positioned at $x = 300$ cells with an initial device spacing of 100m. The hypothetical device porosity structures use the porosity values listed in Table 1.

Reflected and Transmitted Spectra

To assess the device performance the reflected and transmitted spectra were calculated.

<table>
<thead>
<tr>
<th>Frequency</th>
<th>Porosity</th>
</tr>
</thead>
<tbody>
<tr>
<td>0.060</td>
<td>0.844</td>
</tr>
<tr>
<td>0.073</td>
<td>0.814</td>
</tr>
<tr>
<td>0.087</td>
<td>0.804</td>
</tr>
<tr>
<td>1.000</td>
<td>0.806</td>
</tr>
<tr>
<td>0.113</td>
<td>0.811</td>
</tr>
<tr>
<td>0.127</td>
<td>0.820</td>
</tr>
<tr>
<td>0.140</td>
<td>0.832</td>
</tr>
<tr>
<td>0.153</td>
<td>0.841</td>
</tr>
<tr>
<td>0.167</td>
<td>0.841</td>
</tr>
<tr>
<td>0.180</td>
<td>0.879</td>
</tr>
<tr>
<td>0.193</td>
<td>0.871</td>
</tr>
<tr>
<td>0.207</td>
<td>0.890</td>
</tr>
<tr>
<td>0.220</td>
<td>0.880</td>
</tr>
<tr>
<td>0.233</td>
<td>0.884</td>
</tr>
<tr>
<td>0.247</td>
<td>0.887</td>
</tr>
</tbody>
</table>

The reflected spectrum was calculated using a two wave gauge approach as defined in (Goda, Suzuki 1976). The use of multiple wave height measurements with a separation of 40m was used to calculate $K_R$ (reflection coefficient) at each frequency interval. The frequency varying $K_R$ (shown in Fig. 2) yields a similar distribution to that presented in Fig. 1, where a peak $K_R$ value of above 0.11 occurs at 0.1Hz. At higher frequencies (>0.2Hz) an increase in reflection is depicted, which may be a result of the wave gauge separation nearing the recommended limit stated in (Goda 1985). The data from the reflected spectrum shows increased device reflection at the peak frequency.

![Fig. 1 De-normalized frequency dependent power capture coefficient based on experimental data. 15 frequency intervals have been plotted on the smoothed experimental data.](image-url)
The transmitted wave spectrum was calculated by using standard linear wave theory to convert the outputted wave heights behind the porosity device to equivalent energies and hence power spectral densities. The results are shown in Fig. 3, where the undisturbed and transmitted wave spectrum indicates a large reduction around the high power capture frequencies (see Fig. 1). The reduction of 5.9m$^2$/Hz at the 0.1Hz spectral peak represents greatest proportion of wave energy absorption. As the frequency moves away from 0.1Hz the change in PSD between the incident and transmitted spectrum reduce.

Spatial Distance around a Flat Domain

The spatial disturbance around the array was presented in terms of percentage energy change. This uses the summed spectral moments for all frequency components for an energy proportionate wave height. The wave amplitude is then used to calculate energy and the percentage change in energy is then deduced. The results can split into 2 regions, the device up-wave, where wave-device reflections are dominant and the device down-wave, where transmitted effects are dominant.

Care was taken to remove unphysical artefacts resulting from domain sidewall reflections. A region, indicated by a box in Fig 4, was kept clear of such reflections using a wave-speed dependent post-processing tool. For each frequency component, the tool calculates a wave speed, and hence a maximum simulation time before sidewall reflections enter the area of interest. This allows the slower (shorter wavelength) waves to fully propagate across the domain.

The up-wave results show a complex interaction with a highly variable change in energy across the domain. There are several large increases in wave energy exceeding 15%. Down wave, the change in percentage energy reduces as the distance increases. The edges of the domain shows a small area, 150m behind the central device, where an energy increase is observed. This is a result of the device wake as it propagates down-wave symmetrically from each device, when these regions meet a large positive change in energy occurs.

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The up-wave results show the devices located at $x = 650\text{m}$, $750\text{m}$ and $850\text{m}$. Directly in front of each device a large narrow peak is observed, as the distances increases the magnitude severely lessens and the peaks broaden. The complex up-wave energy disturbance further indicates the high spatial variance, where very small correlations between transects can be observed. The $50\text{m}$ and $100\text{m}$ transects are both located near a peak location, the larger distance from the $100\text{m}$ transect shows a much broader spread in disturbance resulting in a low magnitude peak. As the distance increases to $200\text{m}$ the transect shows a large variation at either end, it is this disturbance that if not carefully monitored results in severe sidewall reflections.

The down-wave transects show the device locations clearly indicated by the maximum percentage change in the $0\text{m}$ transect (see Fig. 7). The $0\text{m}$ transect experiences a large narrow absorption behind each device, in between which a slight increases in energy is observed. As the distance increases the magnitude of the energy change reduces and the wake distribution spreads. The $50\text{m}$ and the $100\text{m}$ transect crosses in front and behind the main increase in energy change. This results in a minimal energy change behind the devices. The individual device disturbances decrease as the down-wave distance increases, this is illustrated in the $200\text{m}$ transects, where a broad almost U-shaped curve is shown.

The flat domain test demonstrates the application of the frequency dependant absorption. The analysis of the device spectrum shows the presences of the peak reflection and transmission occurring at the peak power capture of the device. The device transects provides large amounts of data on the individual device disturbance interactions. This shows regions of large positive and negative energy changes with several hundred metres of the array.
1350m respectively, where the device is located at 600m. The up-wave near device disturbance is shown to be identical between tests. As the initial peak begins to oscillate a small increase in disturbance is shown at the peaks and troughs. When the distance in front of the device increases, a slight difference between the wavelengths of the disturbance oscillations can be seen. This is also shown with 1:50 test when the disturbance wavelength is much closer to that of the 1:100 results. The down-wave results exhibit identical disturbances directly behind the array. As the distance increases, combined wake effects produce a peak, the magnitude of which is shown to be slightly larger for the steeper gradient tests. Beyond this a similar disturbance level is observed for all tests. The very nearshore data (depths = 2-3m) are not presented in Fig. 9 as they are located within the end sponge layer and therefore are subject to varying levels of dissipation.

CONCLUSIONS

This study has demonstrated the method of implementing a frequency dependent WEC simulation within the MIKE21 Boussinesq wave model. With the use of experimental frequency distributed power capture a series of nearshore OWSCs was simulated. The immediate device data shows a peak reflection coefficient of over 0.11 at the peak frequency. An absorption of 5.9m²/Hz was observed immediately behind a device. The spatial results indicate a complex reflection pattern that causes varying degrees of wave disturbance over short distances. The down-wave results show a general reduction to the change in energy as the distance behind the devices increases. As wake effects interact with these disturbances causing regions of increased energy resulting in a neutral energy change behind the device.

Sloped gradient tests revealed that an increased domain angle caused large changes in the spatial up-wave disturbances, where the steeper slope showed an increase in peak disturbance. The down-wave results reveal a reduced dependency on slope angle; however, wake interaction provided a noticeably larger disturbance for the steep domain. Interestingly, these findings demonstrate that bathymetric slope has an important role on reflected wave-device interaction but only limited effects on the down-wave results.

In recent years, Boussinesq models have often been adopted as a promising pragmatic approach to the challenging problem of incorporating the effects of WECs in wave simulations. This study provides an important refinement to the existing methodology, by developing a mechanism to incorporate frequency-dependent absorption. The next challenge for such models is an explicit treatment of waves radiated from devices. While the effect of radiation will be minimal on the mid-field region of interest in this study, it will be more important in the near-field, and could be simulated using an additional WGL. However, as with most other models, there are restrictions on the availability of phase information for device emitted radiation, which will be strongly dependent on the design, configuration and tuning of the device. In addition, validation of this modelling technique using experimental data would be highly desirable.

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REFERENCES


